# **Fault Detection and Estimation of a Lithium-Ion Battery System Using an Adaptive Observer**

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## **ABSTRACT**

Nowadays, we are dealing with the increasing complexity of industrial systems, which are often equipped with a large number of sensors and actuators. Industrial processes are usually complex and consequently vulnerable. The likelihood of multiple failures and resulting economic losses also increases. Therefore, fault estimation is gaining more and more attention from a practical point of view and is an important aspect in modern fault diagnosis (FD), which can provide knowledge about the detection, isolation and identification of the faults. In this paper a novel fault detection and estimation adaptive based observer approach for the Takagi-Sugeno (T-S) system is proposed.

**Keywords:** Fault diagnosis, Fault detection, Fault estimation, Adaptive observer

## **INTRODUCTION**

Interruption of a process or a poor quality of the final product resulting from a system failure can cause serious economic losses, which are often many times greater than the costs of maintains of industrial system process and components (Jasiulewicz-Kaczmarek 2017, Jasiulewicz-Kaczmarek 2022). Moreover, with the onset of the Industry 4.0 era (Waszkowski 2022) and the increase in the degree of automation, robotisation and computerisation (Kiedrowicz 2016a, Kiedrowicz 2016b, Waszkowski 2016), there is a further increase in the number of system components, sensors and actuators. Such situation may increase the likelihood of simultaneous failure of various system components. Moreover, the high complexity of modern systems requires special support for the operators who manage them. The abundance of diagnostic information may cause decision-making errors, especially in the area of system control, which may lead to significant material losses. The design of modern technical systems often aims to either limit the impact of the human factor on a system/process or to design the system in such a way that it provides necessary and reliable information enabling decision-making by the system operator. This article develops a fault estimation method that provides the operator precise information about the moment of occurrence and characteristics of fault. Fault detection and identification (FDI) (Chen 2017, Lan 2016, Mrugalski 2007) methods for fault diagnosis include standard

fault detection and location procedures, while fault estimation methods are additionally used to estimate the signals of the faults. In the literature, it can often find a combination of FDI and fault estimation (FE) (Zhirabok 2015, Rodrigues 2015) methods due to the fact that FDI schemes alone do not provide estimation of real fault. In this case, we are dealing with a two-stage operation, where first the detection and location of fault takes place, and then the stage of fault estimation begins, i.e., the estimation of its course in time. Despite many publications on the problem of fault estimation, it seems that this issue is still an open research problem. Most research publications is focused on estimating only one type of fault, i.e., the actuator (Lan 2016, Pazera 2020) or the sensor (Chen 2017, Zhang 2017, Zhang 2018). Although in the current literature there are results of research on observers that are able to detect simultaneous faults of actuators and sensors, the disadvantage of most of them is the assumption of a limited error rate, i.e., it is assumed that the error derivative is close to zero. Nevertheless, among the existing fault estimation strategies, those based on sliding observers (Xia 2017, Feng 2020) and the Kalman filter (Gou 2018) are of key interest. Considering the above, it is natural that the problem of simultaneous estimation of actuator and sensor faults is gaining more and more attention of the authors of numerous research papers (Liu 2018, Youssef 2017, Lan 2016, Chaves 2019). Nevertheless, these methods are mostly designed for linear systems, while the number of strategies capable of handling some classes of non-linear systems is rather limited. In order to solve this problem, a procedure for designing an adaptive observer for the Takagi-Sugeno system was developed in this paper. The Takagi-Sugeno (Takagi 1985) system takes into account the possibility of fault to the actuator and sensor, as well as possible disturbances in the form of measurement and process uncertainty. The main feature of the proposed adaptive observer is its simple design procedure, where it consists of three separate estimations of the state, sensor and actuator faults with the application of its own gain matrices. As a consequence, the procedure for designing an adaptive observer boils down to the calculation of a linear matrix inequality and the calculation of three reinforcement matrices. The advantage of the developed observer is the simple implementation. Finally, the correctness and accuracy of the developed observer was validated in the work using a battery system whose Takagi-Sugeno model was developed on the basis of experimental data.

#### **ADAPTIVE OBSERVER FOR T-S SYSTEMS**

The system in the form of T-S, taking into account the occurrence of fault and disturbances, can be defined as follows:

$$
x_{f,k+1} = A (p_k) x_{f,k} + B (p_k) u_{f,k} + B (p_k) f_{a,k} + W_1 w_{1,k}
$$
  
= 
$$
\sum_{i=1}^{M} h_i (p_k) [A^i x_{f,k} + B^i (p_k) u_{f,k} + B^i (p_k) f_{a,k}] + W_1 w_{1,k},
$$
 (1)

$$
y_{f,k} = Cx_{f,k} + C_f f_{s,k} + W_2 w_{2,k},
$$
 (2)

and

$$
b_i(p_k) \ge 0, \quad \forall i = 1, ..., M_m, \quad \sum_{i=1}^{M} b_i(p_k) = 1,
$$
 (3)

where  $x_{f,k} \in X \subset \mathbb{R}^n$ ,  $u_{f,k} \in \mathbb{R}^r$  and  $y_{f,k} \in \mathbb{R}^m$  are the state, input, and output vectors, respectively. The A, B, and C matrices are known system state, input, and output matrices, where  $W_m$  indicates the number of T-S models. Also, the symbols  $f_{a,k} \in F_a \subset R^{n_a}$  and  $f_{s,k} \in F_s \subset R^{n_s}$  denote actuator and sensor fault vectors. In addition, the  $C_f$  matrix points to the sensor failure distribution matrix, where rank( $C_f$ ) =  $n_s$  and  $n_a + n_s \leq m$ . This means that it is possible to estimate faults of the sensors and actuators only when their number does not exceed the number of measurable system outputs. Whereas,  $W_1$  and  $W_2$  are vector distribution matrices  $w_{1,k}$  and  $w_{2,k}$ , which represent external disturbances in the form of measurement uncertainty and process. Finally, the following notation is used throughout the rest of the subsection

$$
A(p_k) = \sum_{i=1}^{M} b_i(p_k) A^i,
$$
 (4)

In this paper, the following equations describing the state estimates, faults of actuators and sensors have been proposed:

$$
\widehat{\mathbf{x}}_{f,k+1} = A\left(p_k\right)\widehat{\mathbf{x}}_{f,k} + B\left(p_k\right)\mathbf{u}_{f,k} + B\left(p_k\right)\widehat{f}_{a,k} ++ \mathbf{W}_1\mathbf{w}_{1,k} + \mathbf{K}_x\left(p_k\right)\left(\mathbf{y}_{f,k} - \mathbf{C}\widehat{\mathbf{x}}_{f,k} - \mathbf{C}_f\widehat{\mathbf{f}}_{s,k}\right),\tag{5}
$$

$$
\widehat{f}_{a,k+1} = \widehat{f}_{a,k} + K_a \left( p_k \right) \left( y_{f,k} - C \widehat{x}_{f,k} - C_f \widehat{f}_{s,k} \right), \tag{6}
$$

$$
\widehat{f}_{s,k+1} = \widehat{f}_{s,k} + K_s (p_k) \left( y_{f,k} - C \widehat{x}_{f,k} - C_f \widehat{f}_{s,k} \right). \tag{7}
$$

The state estimation error can be described in the following concise form:

$$
\overline{e}_{f,k+1} = \widetilde{A} (p_k) \overline{e}_{f,k} + \widetilde{W} (p_k) \overline{w}_k, \qquad (8)
$$

where:

$$
\widetilde{A} (p_k) = \overline{A} (p_k) - K_f (p_k) \overline{C}, \quad \overline{A} (p_k) = \begin{bmatrix} A (p_k) & B (p_k) & 0 \\ 0 & I & 0 \\ 0 & 0 & I \end{bmatrix},
$$
\n
$$
K_x (p_k) = \begin{bmatrix} K_x^T (p_k) & K_a^T (p_k) & K_s^T (p_k) \end{bmatrix}^T, \quad \widetilde{W} (p_k) = \overline{W}_3 - K_f (p_k) \widetilde{W}_2,
$$
\n
$$
\overline{W}_3 = \begin{bmatrix} W_1 & 0 & 0 & 0 \\ 0 & 0 & I & 0 \\ 0 & 0 & 0 & I \end{bmatrix}, \quad \widetilde{W}_2 = \begin{bmatrix} 0 & W_2 & 0 & 0 \end{bmatrix}, \quad \overline{C} = \begin{bmatrix} C & 0 & C_f \end{bmatrix}.
$$

Therefore, the following theorem is proposed for the design of the observer: The system based on observer 8 is strictly quadratic bounded for

all values  $\tilde{w}_k \in E_w$  if there are matrices  $P_f > 0$  and  $N_f(p_k)$  as well as the value as well as the value  $\alpha \in (0, 1)$  such that the following condition is met:

$$
\begin{bmatrix}\n-P_f + \alpha P_f & * & * \\
0 & -\alpha Q_w & * \\
P_f \overline{A} (p_k) - N_f (p_k) \overline{C} & P_f \overline{W}_3 - N_f (p_k) \overline{W}_2 & -P_f\n\end{bmatrix} < 0.
$$
\n(9)

Finally, the procedure for designing the T-S adaptive observer developed in this chapter boils down to calculating the linear matrix inequality (9) and obtaining the following reinforcement matrices:

$$
K_f(p_k) = \begin{bmatrix} K_x(p_k) \\ K_a(p_k) \\ K_s(p_k) \end{bmatrix} = P_f^{-1} - N_f(p_k). \qquad (10)
$$

### **AN EXAMPLE OF FAULT ESTIMATION OF A LITHIUM-ION BATTERY SYSTEMS**

The purpose of this section is to present the efficiency of the developed adaptive observer on the example of state and faults estimation of the lithium-ion battery system. To achieve this goal the following equivalent second-order resistor-capacitor circuit model is considered:

$$
\begin{cases}\n\dot{U}_1 = -\frac{U_1}{R_1 C_1} + \frac{I_b}{C_1} \\
\dot{U}_2 = -\frac{U_2}{R_2 C_2} + \frac{I_b}{C_2}, \\
\dot{SOC} = \frac{I_b}{3600 C_b}\n\end{cases} (11)
$$

where  $U_1, R_1, C_1$  and  $U_2, R_2, C_2$  are the voltage, resistance and capacitance variables of the first and second resistor-capacitor circuits, respectively. Moreover, the value  $I_b$  represents the system current and  $C_b$  represents the nominal capacity, which is  $C_b \approx 2[Ah]$ . A graphical diagram of the resistor-capacitor circuit is shown in the Figure 1.



**Figure 1:** A graphical diagram of the second-order resistor-capacitor circuit.

The following faults scheme was proposed to verify the correctness and accuracy of the developed observer:

$$
f_{a,k} = \begin{cases}\n-0,35 \cdot u_k & 7000 \le k \le 9000 \\
a \cdot e^{\text{bj}} & 11000 \le k \le 14000 \\
j = 1,2,...,3000 & 0,13 \cdot u_k & 19000 \le k \le 22442 \\
0 & 0,13 \cdot u_k & 19000 \le k \le 22442 \\
0 & \text{in other case}\n\end{cases}
$$
\n
$$
f_{s,k} = \begin{cases}\ny_{f,k} + 0,013 & 4500 \le k \le 7500 \\
y_{f,k} - j \cdot 4 \cdot 10^{-4} & 10000 \le k \le 15000 \\
where & j = 1,2,...,5000 \\
0 & 19000 \le k \le 22442 \\
0 & \text{in other case}\n\end{cases}
$$
\n(13)

along with the fault distribution matrix of current and voltage measurement:

$$
\mathbf{B}_{f} = \frac{1}{M_{m}} \sum_{i=1}^{M_{m}} \mathbf{B}_{f}^{i} = \begin{bmatrix} 1.856086 \\ 0.4853711 \\ 0.1389095 \end{bmatrix} \cdot 10^{-3}, \quad \mathbf{C}_{f} = \begin{bmatrix} 0 \\ 1 \end{bmatrix}.
$$
 (14)

Note that the voltage measurement fault distribution matrix contains either zeros or ones. One means that the fault affects the given measurement, and zero means the opposite.

Such a faults scenario allows to check the correctness and performance of the proposed observers in four cases:

- Instantaneous permanent fault;
- Linearly increasing fault for voltage measurement  $f_{s,k}$ ;
- Exponential fault for current measurement  $f_{a,k}$ ;
- Permanent fault of a fixed value.

Moreover, the disturbance distribution matrices in the form of measurement and process uncertainty are defined as follows:

$$
W_1 = 1 \cdot 10^{-6} I, \quad W_2 = 1 \cdot 10^{-2} I. \tag{15}
$$

It should be noted that the interference distribution matrices were defined using experimental data and the general approach proposed in (Witczak 2015). In the proposed case, the failure of the voltage measurement  ${f}_{s,k}$  corresponds to possible significant measurement inaccuracies. Similarly, the failure of the current measurement  $f_{a,k}$  corresponds to a significant loss of battery performance related to its behaviour based on the discharge current. In addition, you can see that positive and negative fault values are included in the faults scenario. Moreover, it is clear that the voltage measurement fault distribution matrix  $C_f$  is defined on the basis of the C matrix of the (11) model. Furthermore, from the  $C_f$  matrix, it should be noted that the  $f_{s,k}$  corruption

occurred in the  $U_2$  state of the system. Moreover, on the basis of the equations (12)-(13) it is worth stating that the faults of the voltage and current measurement occurred simultaneously in the given moments of time.

Figures 2–3 show, respectively, voltage  $f_{s,k}$  and current  $f_{a,k}$  faults according to the fault scenario (12)-(13). The blue line shows the actual fault, while the red line shows its estimation. It can be seen that the quality of faults estimation is very good in both cases, despite the disturbances. In addition, an important aspect is that in both cases there are three types of fault, i.e., temporary, permanent and slowly increasing. In addition, it should be noted that at given moments of time, voltage and current measurement failures occur simultaneously, which is an additional difficulty in estimating their value.



**Figure 2:** Voltage measurement fault.



**Figure 3:** Current measurement fault.

#### **CONCLUSION**

This paper presents the construction of an adaptive observer. The main feature of the adaptive observer is its simple design procedure, where it consists of three separate estimations for estimating the condition and faults of the sensor and actuator, each with its own gain matrix. As a consequence, the procedure for designing an adaptive observer boils down to the calculation of a linear matrix inequality and the calculation of three reinforcement matrices. Therefore, the advantage of the developed observer is the possibility of its simple implementation, however, the control system should have a relatively large computing power due to the number of gain matrixes and additional disturbances resulting from mathematical calculations. The designed observers for estimating the faults of the sensor and actuator provides the operator precise information about the moment of occurrence and characteristics of fault.

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